

Turbo Intercalación en los Sistemas de Comunicaciones Moviles Cdma2000 y W-cdma: Tutorial

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RESUMEN

En este artículo se presenta un análisis detallado de la operación de los intercaladores empleados por los turbo códigos paralelos concatenados definidos en los estándares de telecomunicaciones cdma2000 (3GPP2 C.S0024-B V2.0) y W-CDMA (3GPP TS 25.212 V7.4.0). Se discuten las diferencias en el algoritmo de permutación y se hace un análisis estadístico de dispersión y frecuencia para cada turbo intercalador. Finalmente se presentan dos ejemplos para ilustrar el proceso completo de turbo intercalación definido en cada estándar.

PALABRAS CLAVE: Turbo codificación, cdma2000, W-CDMA, 3G, intercalación, comunicaciones digitales.

Turbo Interleaving Inside the Cdma2000 and W-cdma Mobile Communication Systems: a Tutorial

ABSTRACT

In this paper a detailed analysis on the operation of the interleavers employed by the parallel concatenated turbo codes defined in the telecommunications standards cdma2000 (3GPP2 C.S0024-B V2.0) and W-CDMA (3GPP TS 25.212 V7.4.0) is presented. Differences in the permutation algorithm employed by each turbo interleaver as well a statistical dispersion and frequency analysis for each one are also discussed. Finally, two examples are presented to illustrate the complete turbo interleaving process defined in each standard.

KEYWORDS: Turbo coding, cdma2000, W-CDMA, 3G, interleaving, digital communications.

1. INTRODUCTION

Near one decade ago the IMT2000 initiative of the International Telecommunication Union identified five base station-mobile station air interfaces for the third-generation mobile communications systems (3G). It seems clear at the present however that the two technologies that will be dominating the global market of third generation mobile communications systems are cdma2000 and W-CDMA [1].

The salient feature of third generation mobile communications systems is its high capacity for transmitting information over the system data channels. By 1997 IMT2000 defined in Recommendation ITU-R M.1225 test data rates of 2048 kbit/s, 144 kbit/s and 64 kbit/s for indoors, pedestrian and vehicular traffic respectively [2] for purposes of evaluating the third-generation technologies. As expected, the continuing evolution of mobile technologies has left behind these values with much higher speeds. For instance, the Ultra Mobile Broadband™ (UMB™) air interface specification [3] is intended to deliver downlink and uplink data rates of 288 Mbit/s and 75 Mbit/s respectively using a bandwidth of 20 MHz.

To offer these high data rates with access terminals increasingly both small and functional it is imperative to work at the limit of efficiency in data transmission.

As it is well known in 1948 C. E. Shannon proved that the fundamental limit of digital transmission on channels with white noise is given by the classic channel capacity formula $C = W \log_2(1 + S/N)$, where C is the capacity in bit/s, W is the channel bandwidth in Hz, and S/N is the signal to noise ratio at the receiver.

However, finding an error correction system able to achieve this limit meant extensive research for several decades. After more than forty years of research the concept of turbo coding developed by Claude Berrou and Alain Glavieux [4] finally proved that it was possible to reach the limit of channel capacity with an encoding scheme that could be constructed in practice.

While turbo coding is not the only technique known to be able to attain the channel capacity limit [5] it is certainly the most commonly used channel coding technique for data channels in contemporary mobile communications systems.

According to the inventors the turbo coding principle was born from the experimentation with the feedback concept applied to the error correcting problem using convolutional codes [6]. At the core of a turbo coding system there is a fundamental constitutive element called interleaver. An interleaver is a system that changes the positions of input data according to an established position permutation algorithm. Inside the turbo coding process the function of the interleaving block is to help in providing codes vectors with the highest possible level of randomness (ideally, independent vectors) [7] so that the resulting code resembles as close as possible the concept of random coding used by C. E. Shannon in [8] to prove the channel capacity theorem. Therefore the interleaver is a fundamental element for the performance of a turbo code [9].

The understanding of interleaving is a subject of high interest to the specification of physical layers for both wired and wireless transmission technologies. The aim of this article is to show in detail how the turbo interleavers defined in the cdma2000 EV-DO Revision B (3GPP2 C.S0024-B V2.0) [10] and W-CDMA (3GPP TS 25.212 V7.6.0) [11] standards work, what are their main characteristics, and what are the design principles used by each one. This article has been written with a tutorial approach in mind.

The article is organized as follows: Section 2 provides an overview of the turbo interleavers defined in cdma2000 and W-CDMA standards. In Section 3 two detailed examples of the turbo interleaving done by each standard are described. Section 4 shows a dispersion analysis for each interleaver. Finally, in section 5 the main findings and observations of this work are presented.

For a full discussion of the theory of the interleavers described in this article references [12] and [13] can be consulted.

2. MATERIALS AND METHODS

W-CDMA and cdma2000 use different strategies for the interleaving carried out by its turbo coding systems. The cdma2000 interleaver is based on the principle of generating the interleaving positions through a counter that generates addresses which are modified through a preset table and a function that reverses the order of the bits. The resulting address vectors determine the permutation of the input data. In cdma2000 the input to the interleaver and the output data form the interleaver are defined as arrays (vectors) of length N_{turbo} . The values that the N_{turbo} variable can take are defined by the standard.

In W-CDMA the input and output of the interleaver are treated as matrices whose dimensions (rows and columns) depend on the total length of the input data, K . The values the variable K can take are also defined by the standard. The interleaving process takes place in two steps. First the positions of the bits for each row are permuted. Then the row positions are permuted (without changing the bits in each row). In summary, the American standard (cdma2000) treats input bits as an array (or vector), while the European standard (W-CDMA) treats the input data as a matrix. To change positions the American standard uses a counter while the European one uses permutation patterns of rows and columns. The following sub-sections describe in detail the operation of the cdma2000 and W-CDMA interleavers.

2.1 cdma2000 turbo interleaver

Figure 1 shows the flow diagram of the interleaver used by the cdma2000 turbo encoder, which has as input the `packet_size` variable which is used to determine from Table 1 both then n y N_{turbo} parameters. The value of n is a interleaving parameter defined as an integer in the range $3 \leq n \leq 7$. N_{turbo} is the actual number of information bits in the interleaving block and must satisfy the relationship $N_{turbo} \geq 2^{n+5}$.

The packet size is six bit longer than N_{turbo} because the six trail bits are used to force the turbo encoder to the initial state after the codification of the N_{turbo} data bits is complete.

The parameters in Table 1 are defined for reverse link channels i.e. channels going from the mobile station to the base station. For forward link channels the (base station to mobile) n is in the range

Table 1. Turbo interleaver Parameter

<i>Packet_size</i>	<i>n</i>	<i>N_turbo</i>
256	3	250
512	4	506
1024	5	1018
2048	6	2042
4096	7	4090

Figure 1 shows the sequential tasks that are performed within the interleaver proposed in the cdma2000 standard. The first task is to calculate the *MSB* address by taking the n least significant bits of the value of the address counter n most significant bits plus one.

Then Table 2 is indexed using the counter's five least significant bits. This lookup table indexing provides an *LSB* address of n bits.

The next step is to take the n least significant bits of the product of the previously obtained *MSB* and *LSB* addresses which will constitute the lower part, namely *LSB_address*, of the final address.

The higher part of the final address, namely *MSB_address*, is obtained by bit-reversing the five least significant bits of the counter.

MSB_address and *LSB_address* are then concatenated forming the final address which is stored at the *Output_address* vector if the final address is less than N_{turbo} , otherwise the address is discarded.

The counter is increased by one and the process is repeated until the N_{turbo} interleaving addresses are obtained.

The algorithm is designed so that with 2^{n+5} iterations is always possible to obtain the N_{turbo} required addresses, i.e. it is not possible for the iterations to end without having obtained all the interleaving addresses.

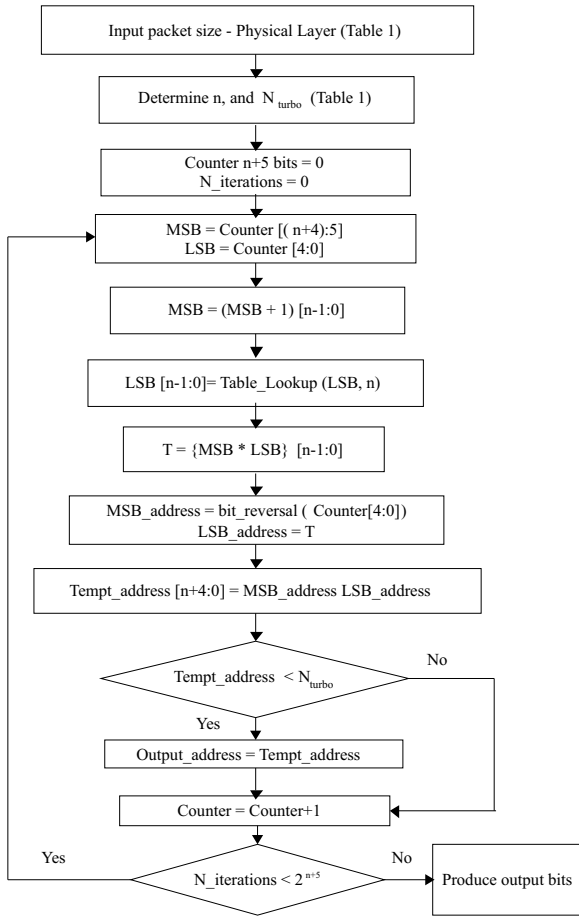


Fig.1. Flow diagram for the cdma2000 standard's turbo interleaver algorithm

2.2 W-CDMA turbo interleaver

The turbo interleaver of the W-CDMA standard is based on a rectangular input bit matrix. This matrix is permuted both by columns and rows before the output bits are delivered. The input bits are denoted as $x_1, x_2, x_3, x_4, x_5, \dots, x_k$ where K is the number of input bits, where $40 \leq K \leq 5114$.

Figure 2 shows a flow diagram of the algorithm used by the W-CDMA turbo interleaver.

As shown in Fig. 2, after verifying that the length of K is in the range established by the standard, then both the number of rows R , and the number of columns C of the rectangular matrix are determined according to the rules given in Tables 3 and 4.

Table 2. cdma2000 turbo interleaver lookup table

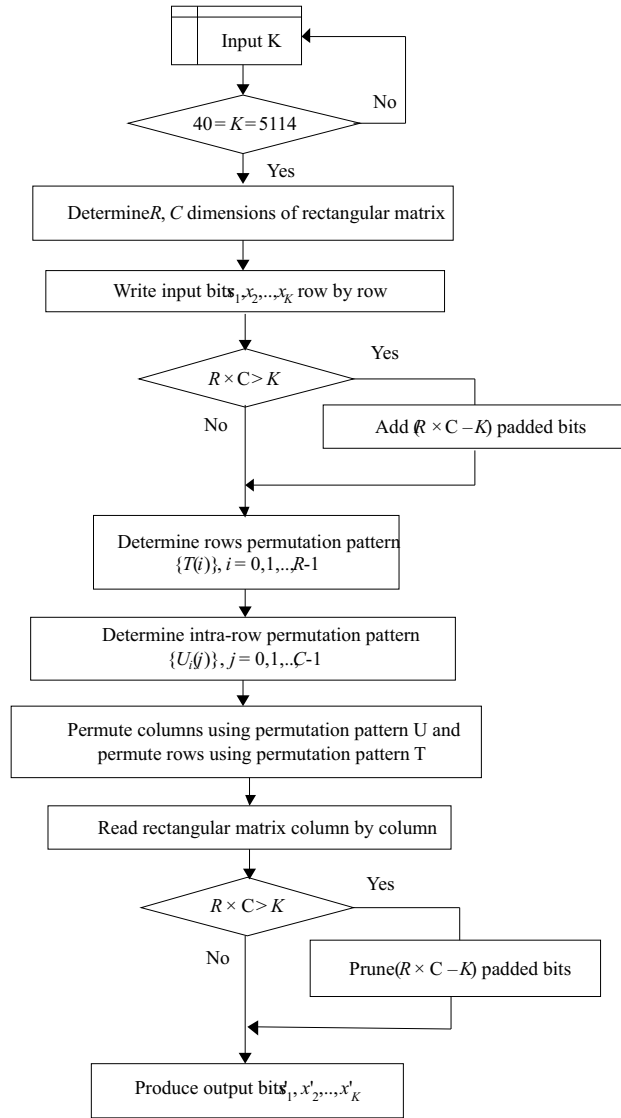
Index	$n=3$	$n=4$	$n=5$	$n=6$	$n=7$
0	1	5	27	3	15
1	1	15	3	27	127
2	3	5	1	15	89
3	5	15	15	13	1
4	1	1	13	29	31
5	5	9	17	5	15
6	1	9	23	1	61
7	5	15	13	31	47
8	3	13	9	3	127
9	5	15	3	9	17
10	3	7	15	15	119
11	5	11	3	31	15
12	3	15	13	17	57
13	5	3	1	5	123
14	5	15	13	39	95
15	1	5	29	1	5
16	3	13	21	19	85
17	5	15	19	27	17
18	3	9	1	15	55
19	5	3	3	13	57
20	3	1	29	45	15
21	5	3	17	5	41
22	5	15	25	33	93
23	5	1	29	15	87
24	1	13	9	13	63
25	5	1	13	9	15
26	1	9	23	15	13
27	5	15	13	31	15
28	3	11	13	17	81
29	5	3	1	5	57
30	5	15	13	15	31
31	3	5	13	33	69

Table 3. Rules for determining the number of rows R

R	K
5	$40 \leq K \leq 159$
10	$(160 \leq K \leq 200) \vee (481 \leq K \leq 530)$
20	$K = \text{any other value}$

Table 4. Rules for determining the number of columns C

K	P	C
$481 \leq K \leq 530$	53	$C = p = 53$
$K \leq R \times (p+1)$	Mini-mum p (see Table 5)	$C = p-1$ if $(K \leq R \times (p-1))$ $C = p$ if $(R \times (p-1) < K \leq R \times p)$ $C = p+1$ if $((R \times p) < K)$



As shown in Table 4 when K is outside the range $481 \leq K \leq 530$, p is the lowest prime number such that $K \equiv R(p+1) \pmod{C}$ and C is calculated according to the third column of Table 4. The values for the prime number p are shown in Table 5.

The matrix is filled with the K input bits by rows from top to bottom. If $R \times C > K$, then the matrix is zero (or one) padded.

Table 5. List of prime number p and primitive roots v

p	v	p	v	p	v	p	v	p	v
7	3	47	5	101	2	157	5	223	3
11	2	53	2	103	5	163	2	227	2
13	2	59	2	107	2	167	5	229	6
17	3	61	2	109	6	173	2	233	3
19	2	67	2	113	3	179	2	239	7
23	5	71	7	127	3	181	2	241	7
29	2	73	5	131	2	191	19	251	6
31	3	79	3	137	3	193	5	257	3
37	2	81	2	139	2	197	2		
41	6	89	3	149	2	199	3		
43	3	97	5	151	6	211	2		

Table 6. Rules for determining the row permutation pattern

K	R	Inter-row permutation patterns $\langle T(i) \rangle_{i \in \{0,1,\dots,R-1\}}$
$40 \leq K \leq 159$	5	$\langle 4,3,2,1,0 \rangle$
$(160 \leq K \leq 200) \vee (481 \leq K \leq 530)$	10	$\langle 9,8,7,6,5,4,3,2,1,0 \rangle$
$(2281 \leq K \leq 2480) \vee (3161 \leq K \leq 3210)$	20	$\langle 19,9,14,4,0,2,5,7,12,18,16,13,17,15,3,1,6,11,8,10 \rangle$
$K = \text{any other value}$	20	$\langle 19,9,14,4,0,2,5,7,12,18,10,8,13,17,3,1,16,6,15,11 \rangle$

The row permutation pattern which is represented by the vector $\langle T(i) \rangle_{i \in \{0,1,\dots,R-1\}}$ and depends on the number of input bits K as showed in Table 6.

The next task is to calculate the permutation pattern for the columns in each row. This pattern is defined by the matrix $\langle U_i(j) \rangle_{j \in \{0,1,\dots,C-1\}}$ as indicated in Table 7.

Condition	Column permutation pattern $\langle U_i(j) \rangle_{j \in \{0,1,\dots,C-1\}}$
$C = p$	$U_i(j) = s((j \times r_i) \bmod (p-1)),$ $j = 0,1,\dots,(p-2), U_i(p-1) = 0$
$C = p+1$	$U_i(j) = s((j \times r_i) \bmod (p-2)),$ $j = 0,1,\dots,(p-2), U_i(p-1) = 0$ and $U_i(p) = p$ $K = R \times C$, exchange $U_{R-1}(p)$ with $U_{R-1}(0)$
$\frac{C}{p-1}$	$U_i(j) = s((j \times r_i) \bmod (p-1)), j = 0,1,\dots,(p-2)$

In Table 7 the variable s corresponds to the base sequence used to generate the column permutation pattern and is defined by the eq. (1):

$$s(j) = s((v \times s(j-1)) \bmod (p)), j=1,2,\dots,(p-2), s(0)=1$$

where v is the primitive root as defined in Table 5. The variable r_i is the sequence of permuted prime integers defined in equation (2):

$$r_{T(i)} = q_i, i = 0,1,\dots,R-1$$

Where the subscript $T(i)$ is the row permutation vector given in Table 6 and q is the sequence of primes determined by the lowest primes q_i such that $g. c. d(q_i, p-1) = 1$ $q_i > 6$ and, $q_i > q_{(i-1)}$, $i = 1,2,\dots,R-1$.

The original input matrix is first column permuted with the column permutation pattern $\langle U_i(j) \rangle_{j \in \{0,1,\dots,C-1\}}$ and then with the row permutation patterns $\langle T(i) \rangle_{i \in \{0,1,\dots,R-1\}}$. Finally data are read by columns from left to right. If extra zero (or one) bits were initially padded these bits must be removed (the number of bits removed equals $(R \times C) - K$, producing the bits $x'_1, x'_2, x'_3, x'_4, x'_5, \dots, x'_k$.

3. RESULTS

3.1 Turbo interleaver standard cdma2000

In this example a packet size $packet_size = 256$ is used. Following the algorithm described in Fig. 1 we have:

1. From Table 1 $n=3$ and $N_{turbo} = 250$.
2. Start the counter of $n+5 = 8$ bits at zero, i.e. counter = 00000000.
3. Save the three most significant bits of counter in the *MSB* variable.
4. Save the five least significant bits of counter in the *LSB* variable.
5. Add 1 to *MSB*, i.e. $MSB = 001$ for the first iteration
6. Use Table 2 with the five *LSBs* of counter and column n ($index = 00000$ for the first iteration) and store the n -bit result in *LSB*, i.e. $LSB = 001$ for the first iteration.
7. Take the n least significant bits of the *MSB* x *LSB* product and store it in T .
8. Bit reverse the five least significant bits of counter and save it in *MSB_address*, i.e. $MSB_address = 00000$ for the first iteration
9. Concatenate *MSB_address* and *LSB_address* = T into *Temp_address*, i.e. $Temp_address = 00000001$ for the first iteration
10. If $Temp_address < N_{turbo}$ then deliver valid *output_address*, i.e. $output_address = 00000001$ for the first iteration
11. Add 1 to counter
12. Go to step 3

Table 8 shows the six addresses which are discarded during the process because of being greater than the value of N_{turbo} .

For this example, the input vector has 250 data, whose values for simplicity are consecutively numbered from one to two hundred and fifty. As the input vector is relatively large in size only some positions with their values are shown in Table 9. Figure 3 shows the interleaved output data vector.

Table 8. Example of invalid interleaving addresses

Address Count MSB LSB	MS B = MS B+1	LSB = Table 2 [LSB	MSB_ address = Rever se	LSB_ address = MSB x	Temp _ address = [MSB]
000 11111	001	011	11111	001 x 011 = 011	11111 011 =
001 11111	010	011	11111	010 x 011 = 110	11111 110 =
011 11111	100	011	11111	100 x 011 = 100	11111 100 =
101 11111	110	011	11111	110 x 011 = 010	11111 010 =
110 11111	111	011	11111	111 x 011 = 101	11111 101 =
100 11111	101	011	11111	101 x 011 = 111	11111 111 =

Table 9. cdma2000 example input vector

1	2	3	...	1	..	24	...	24	24	25
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Fig.3. cdma2000 output data after interleaving

{2 130 68 198 34 166 98 230 20 150 84 214 52 182
118 242 12 142 76 206 44 174 110 238 26 158 90
222 60 190 126 3 131 71 195 35 163 99 227 23 147
87 211 55 179 115 243 15 139 79 203 47 171 107
235 27 155 91 219 63 187 123 4 132 66 200 36 168
100 232 18 152 82 216 50 5 133 69 197 37 165 101
229 21 149 85 213 53 181 117 245 13 141 77 205 45
173 109 237 29 157 93 221 61 189 125 6 134 72 194
38 162 102 226 24 146 88 210 56 178 114 246 16
138 80 202 48 170 106 234 30 154 94 218 64 186
122 7 135 67 199 39 167 103 231 19 151 83 215 51
183 119 247 11 143 75 207 43 175 111 239 31 159
95 223 59 191 127 8 136 70 196 40 164 104 228 22
148 86 212 54 180 116 248 14 140 78 204 46 172
108 236 32 156 96 220 62 188 124 1 129 65 193 33
161 97 225 17 145 81 209 49 177 113 241 9 137 73
201 41 169 105 233 25 153 89 217 57 185 121 249}

As can be seen in Figure 3 the input data has been totally interleaved from their original positions. For instance at position 16 is the element 242 and at position 242 is the element 233, whereas at the same positions in the input vector in Table 9 are the data 16 and 242 respectively.

Figure 4 shows the results of this example plotted in a Cartesian plane, where the x-axis represents the index and the y-axis the position of the output data.

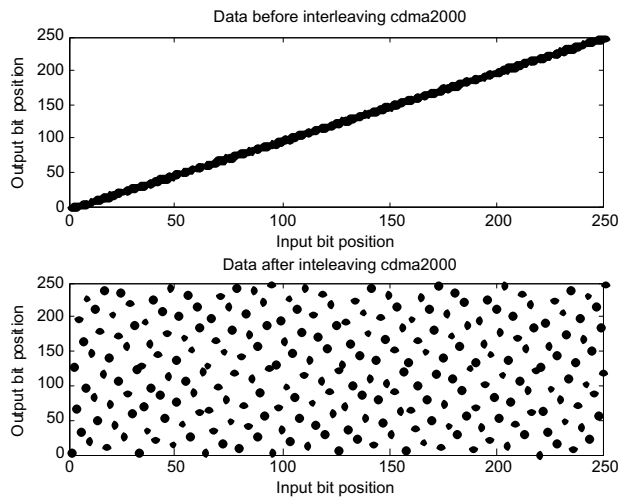


Fig.4. Input vector vs output vector before and after interleaving

3.2 Turbo interleaver standard W-CDMA

For this example a value of $K = 250$ input data is used. From Table 3 the number of rows R must be 20 and from Table 4 the number of columns C must be 13. This value for C is found as follows: from Table 5 the smallest prime such that $250 - 20(p + 1)$ is $p = 13$; according to the rules in Table 4 as $2012 < 2502013$ meets the condition $R(p + 1) < K - R - p$, then $C = p = 13$. The rectangular matrix has then $2013 = 260$ elements. In this example the input values, for simplicity are chosen to be the integers from 1 to 250 which are written by rows into the input matrix. Since the matrix size is greater than K the empty positions are filled with zeros as shown in Table 10.

Table 10. 20x13 matrix before interleaving

1	2	3	4	5	6	7	8	9	10	11	12	13
14	15	16	17	18	19	20	21	22	23	24	25	26
27	28	29	30	31	32	33	34	35	36	37	38	39
40	41	42	43	44	45	46	47	48	49	50	51	52
53	54	55	56	57	58	59	60	61	62	63	64	65
66	67	68	69	70	71	72	73	74	75	76	77	78
79	80	81	82	83	84	85	86	87	88	89	90	91
92	93	94	95	96	97	98	99	10	10	10	10	10
10	10	10	10	10	11	11	11	11	11	11	11	11
11	11	12	12	12	12	12	12	12	12	12	12	13
13	13	13	13	13	13	13	13	13	14	14	14	14
14	14	14	14	14	14	15	15	15	15	15	15	15
15	15	15	16	16	16	16	16	16	16	16	16	16
17	17	17	17	17	17	17	17	17	17	18	18	18
18	18	18	18	18	18	18	19	19	19	19	19	19
19	19	19	19	20	20	20	20	20	20	20	20	20
20	21	21	21	21	21	21	21	21	21	21	22	22
22	22	22	22	22	22	22	22	22	23	23	23	23
23	23	23	23	23	24	24	24	24	24	24	24	24
24	24	25	0	0	0	0	0	0	0	0	0	0

According to the rules in Table 6 the row permutation pattern $T(i)_{i \in \{0,1,\dots,19\}}$ for this case is $T=19,9,14,4,0,2,5,7,12,18,10,8,13,17,3,1,16,6,15,11$. Similarly, as in this example $C=p$, the column permutation patterns according to the rules in Table 7 are determined by equation (3):

$$U_i(j) = s((j \times r_i) \bmod (13-1)),$$

$$j = 0,1,\dots,12, U_i(12) = 0$$

Table 11 shows the base sequence which has been generated using Eq. (1) and which is used in Eq. (3).

Table 11. Base sequence for row permutations

1	2	4	8	3	6	12	11	9	5	10	7
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To obtain the sequence of permuted prime integers r_i the sequence of prime numbers q is needed. The sequence q must satisfy the

condition $g.c.d(q_i, 13-1) = 1, q_i > 6$ and $q_i > q_{i-1}$, $i = 1,2,\dots,19$, in this example giving as a result

$$q = [1,7,11,13,17,19,23,29,31,37,41,43,47,53,59,61,67,71,73,79]$$

The permuted sequence r_i is then:

$$r_{T(i)} = q_i, i = 0,1,\dots,19$$

$$r_{T(i)} = [7,61,19,59,13,23,71,29,43,7,41,79,31,47,11,73,67,53,37,1]$$

Table 12 shows the column permutation patterns produced by Eq. (3).

Table 13 shows the rectangular matrix after applying the permutation pattern U and Table 14 after applying the permutation pattern T . Table 15 shows the resulting output after reading the matrix column by column. It can be observed at the first column of Table 15 the six fill-in zeros.

Figure 5 shows the end result after pruning the fill-in zeros. Figure 6 shows this example's results plotted in a Cartesian plane where the x-axis represents the index and the y-axis the position of the output data.

Table 12. Column permutation matrix $U_i(j)$

2	7	1	9	1	3	1	8	4	6	5	1	1
2	3	5	9	4	7	1	1	1	6	1	8	1
2	1	5	6	4	8	1	3	1	9	1	7	1
2	8	1	6	1	1	1	7	4	9	5	3	1
2	3	5	9	4	7	1	1	1	6	1	8	1
2	8	1	6	1	1	1	7	4	9	5	3	1
2	8	1	6	1	1	1	7	4	9	5	3	1
2	7	1	9	1	3	1	8	4	6	5	1	1
2	1	5	6	4	8	1	3	1	9	1	7	1
2	1	5	6	4	8	1	3	1	9	1	7	1
2	7	1	9	1	3	1	8	4	6	5	1	1
2	1	5	6	4	8	1	3	1	9	1	7	1
2	1	5	6	4	8	1	3	1	9	1	7	1
2	7	1	9	1	3	1	8	4	6	5	1	1
2	8	1	6	1	1	1	7	4	9	5	3	1
2	8	1	6	1	1	1	7	4	9	5	3	1
2	3	5	9	4	7	1	1	1	6	1	8	1
2	1	5	6	4	8	1	3	1	9	1	7	1
2	7	1	9	1	3	1	8	4	6	5	1	1
2	3	5	9	4	7	1	1	1	6	1	8	1
2	3	5	9	4	7	1	1	1	6	1	8	1

Table 13. Rectangular matrix after column interleaving.

2	7	11	9	10	3	13	8	4	6	5	12	1
15	16	18	22	17	20	26	25	23	19	24	21	14
28	38	31	32	30	34	39	29	36	35	37	33	27
41	47	50	45	49	51	52	46	43	48	44	42	40
54	55	57	61	56	59	65	64	62	58	63	60	53
67	73	76	71	75	77	78	72	69	74	70	68	66
80	86	89	84	88	90	91	85	82	87	83	81	79
93	98	102	100	101	94	104	99	95	97	96	103	92
106	116	109	110	108	112	117	107	114	113	115	111	105
119	129	122	123	121	125	130	120	127	126	128	124	118
132	137	141	139	140	133	143	138	134	136	135	142	131
145	155	148	149	147	151	156	146	153	152	154	150	144
158	168	161	162	160	164	169	159	166	165	167	163	157
171	177	180	175	179	181	182	176	173	178	174	172	170
184	190	193	188	192	194	195	189	186	191	187	185	183
197	198	200	204	199	202	208	207	205	201	206	203	196
210	220	213	214	212	216	221	211	218	217	219	215	209
223	228	232	230	231	224	234	229	225	227	226	233	222
236	237	239	243	238	241	247	246	244	240	245	242	235
249	250	0	0	0	0	0	0	0	0	0	0	248

Table 14. Rectangular matrix after both column and row interleaving

249	250	0	0	0	0	0	0	0	0	0	0	248
119	129	122	123	121	125	130	120	127	126	128	124	118
184	190	193	188	192	194	195	189	186	191	187	185	183
54	55	57	61	56	59	65	64	62	58	63	60	53
2	7	11	9	10	3	13	8	4	6	5	12	1
28	38	31	32	30	34	39	29	36	35	37	33	27
67	73	76	71	75	77	78	72	69	74	70	68	66
93	98	102	100	101	94	104	99	95	97	96	103	92
158	168	161	162	160	164	169	159	166	165	167	163	157
236	237	239	243	238	241	247	246	244	240	245	242	235
132	137	141	139	140	133	143	138	134	136	135	142	131
106	116	109	110	108	112	117	107	114	113	115	111	105
171	177	180	175	179	181	182	176	173	178	174	172	170
223	228	232	230	231	224	234	229	225	227	226	233	222
41	47	50	45	49	51	52	46	43	48	44	42	40
15	16	18	22	17	20	26	25	23	19	24	21	14
210	220	213	214	212	216	221	211	218	217	219	215	209
80	86	89	84	88	90	91	85	82	87	83	81	79
197	198	200	204	199	202	208	207	205	201	206	203	196
145	155	148	149	147	151	156	146	153	152	154	150	144

Table 15. Rectangular matrix after reading column by column

249	119	184	54	15	210	80	197	145
250	129	190	55	16	220	86	198	155
0	122	193	57	18	213	89	200	148
0	123	188	61	22	214	84	204	149
0	121	192	56	17	212	88	199	147
0	125	94	59	20	216	90	202	151
0	130	195	65	26	221	91	208	156
0	120	189	64	25	211	85	207	146
0	127	186	62	23	218	82	205	153
0	126	191	58	19	217	87	201	152
0	128	187	63	24	219	83	206	154
0	124	185	60	21	215	81	203	150
248	118	183	53	14	209	79	196	144

In Table 16 and 17 it can easily be seen that the input data are interleaved in a different way by the turbo interleaver of each standard. For instance, in W-DCMA it appears several times consecutive ones at the output (nibbles B16 and C16, for example) but they never come from consecutive positions at the input vector. In cdma2000 it never appears for this example consecutive ones at the output vector.

4. DISCUSSION AND CONCLUSIONS

Although the output produced by interleavers (Figures 3 and 5) seem to have elements of randomness, it should be noticed that the transformation $T[x] = y$ between input and output positions is both deterministic and bijective, i.e. the mapping is fixed, one to one and for all y there is a single x and vice versa.

Because of this reason it is more useful to calculate the average interleaving distance, defined as the average value of the distance between the input and output positions, L_{avg} , as well the standard deviation of these distances.

Table 18 shows the values of L_{avg} and standard deviation for different lengths of input data. As shown in Table 18 the values of L_{avg} and standard deviation are very similar for the two compared standards.

Tabla 18. L_{avg} distance of interleaving

Length of input data	L_{avg}		Standard deviation	
	cdma2000 0	W-CDM A	cdma2000 0	W-CDM A
250	82.19	81.54	58.36	59.23
506	168.96	170.78	118.63	121.65
1018	334.60	333.27	240.71	240.58
2042	678.2	679.68	481.04	483.78
4090	1369.9	1358.8	965.65	964.15

Similarly, as noted at the histograms of interleaving distance versus frequency of Figure 7 and 8 for 250 data, the dispersion patterns are statistically similar. Patterns obtained for higher input values have similar behavior.

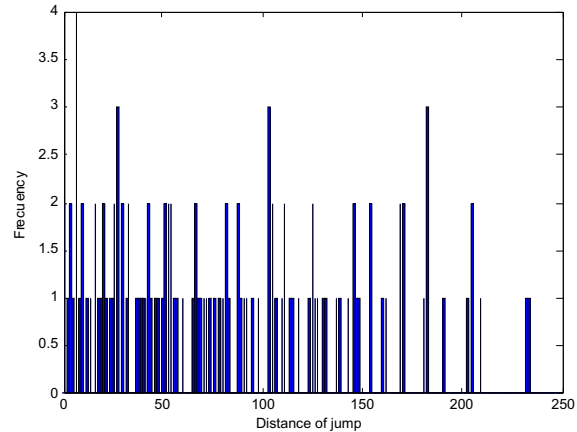


Fig.7. Interleaving distance versus frequency for W-CDMA, packet size = 250.

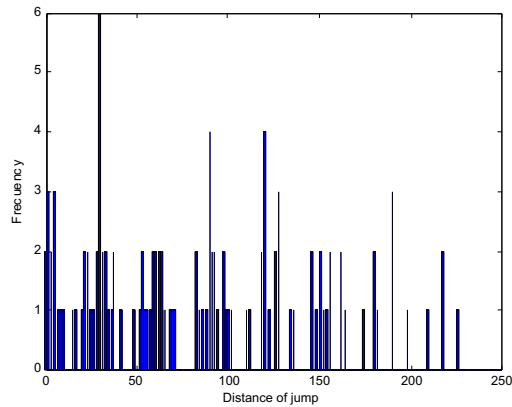


Fig.8. Interleaving distance versus frequency for cdma2000, packet size = 250.

The interleaving process does not mean that all positions must change at the output. For example, in cdma2000 for an input size equal to 506 positions 68, 84 and 338 appear at the same positions at the interleaver's output. For W-CDMA for the same input size (506) it happens the same with position 385.

The main conclusions of this work are:

- In essence the principle of interleaving used in cdma2000 is based on manipulating a counter

whose value defines the mapping position at the output. Instead, the turbo interleaver used by the W-CDMA standard is based on a permutation method using prime numbers for generating permutations for both rows and columns in a rectangular matrix.

- While output vectors appear to be randomly distributed, the transformation which defines the mapping between input and output positions in both W-CDMA and cdma2000 turbo interleavers is bijective and of deterministic nature, i.e. mapping between input and output is fixed.
- Considering the interleaving average distance and the variance of interleaving distance (Table 18) it can be observed that the W-CDMA and cdma2000 turbo interleavers have a very similar behavior even when their interleaving algorithms are substantially different.

The distribution of distances, as can be seen at the histograms of Fig. 7 and Fig. 8, corroborate that the dispersion patterns of the W-CDMA and cdma2000 turbo interleavers are statistically quite similar.

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